

Microwave Resonators

1. Applications

- Oscillators
- Filters
- Sensors

2. Types

- Transmission line resonator: strip, ring, patch (circula, rectangular)
- Cavity resonator
- Dielectric resonator

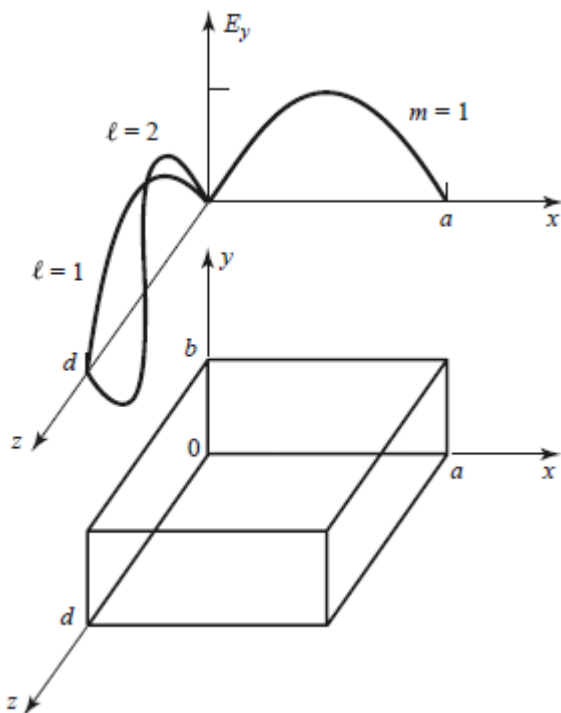
3. Resonator specifications and measurements

- Resonant frequency
- Q factor: unloaded, loaded
- Coupling coefficient

4. Resonator design

- Theory
- Simulation

5. Rectangular cavity resonator



Field expansion:

$$E_x = E_{x0} \cos(k_x x) \sin(k_y y) \sin(k_z z)$$

$$E_y = E_{y0} \sin(k_x x) \cos(k_y y) \sin(k_z z)$$

$$E_z = E_{z0} \sin(k_x x) \sin(k_y y) \cos(k_z z)$$

$$(\nabla^2 + k^2)\mathbf{E} = 0 \rightarrow -k_x^2 - k_y^2 - k_z^2 + k^2 = 0 \quad (k^2 = \omega^2 \mu \epsilon)$$

$$\nabla \cdot \mathbf{E} = 0 \rightarrow k_x E_{x0} + k_y E_{y0} + k_z E_{z0} = 0$$

$$\mathbf{H} = -\frac{1}{j\omega\mu} \nabla \times \mathbf{E}$$

Boundary conditions:

$$k_x = \frac{m\pi}{a}, k_y = \frac{n\pi}{b}, k_z = \frac{p\pi}{d} \quad (m, n, p = 0, 1, 2, \dots)$$

$$k_{mnp}^2 = \omega_{mnp}^2 \mu \epsilon = k_x^2 + k_y^2 + k_z^2 \quad (\text{resonant wavenumber})$$

Lowest-order resonance:

$(m, n, p): (1, 0, 0)$ or $(0, 1, 0)$ or $(0, 0, 1)$ all impossible solutions

$m = 0$ or $n = 0$ or $p = 0$

Select 0 for the smallest dimension among a , b , and d .

TE₁₀₁ mode:

$$E_y = E_0 \sin \frac{\pi x}{a} \sin \frac{\pi z}{d}, \quad E_x = 0, \quad E_z = 0$$

$$H_y = 0$$

$$H_x = \frac{-jk_z}{\omega\mu} E_0 \sin \frac{\pi x}{a} \cos \frac{\pi z}{d}$$

$$H_z = \frac{jk_x}{\omega\mu} E_0 \cos \frac{\pi x}{a} \sin \frac{\pi z}{d}$$

Stored energy

$$W_e = \int_V \frac{1}{4} \epsilon' E^2 dV = \frac{1}{16} \epsilon' E_0^2 abd$$

$$W_m = \int_V \frac{1}{4} \mu' H^2 dV = \frac{1}{16} \epsilon' E_0^2 abd$$

$$W = W_e + W_m = \frac{1}{8} \epsilon E_0^2 abd$$

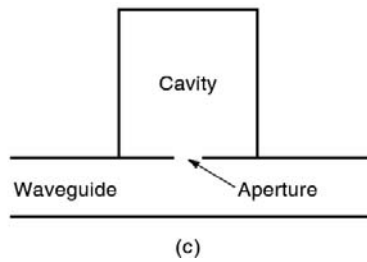
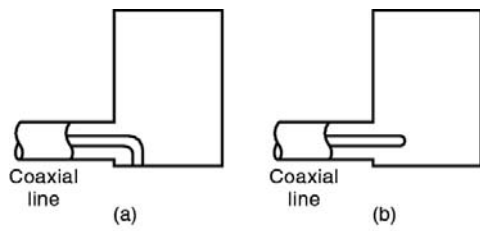
Quality factors:

$$P_c = \frac{R_s}{2} \int_{\text{walls}} |H_t|^2 dS = \frac{R_s E_0^2 \lambda^2}{8(\mu / \epsilon)} \left(\frac{ab}{d^2} + \frac{bd}{a^2} + \frac{a}{2d} + \frac{d}{2a} \right) \quad (\text{conductor loss})$$

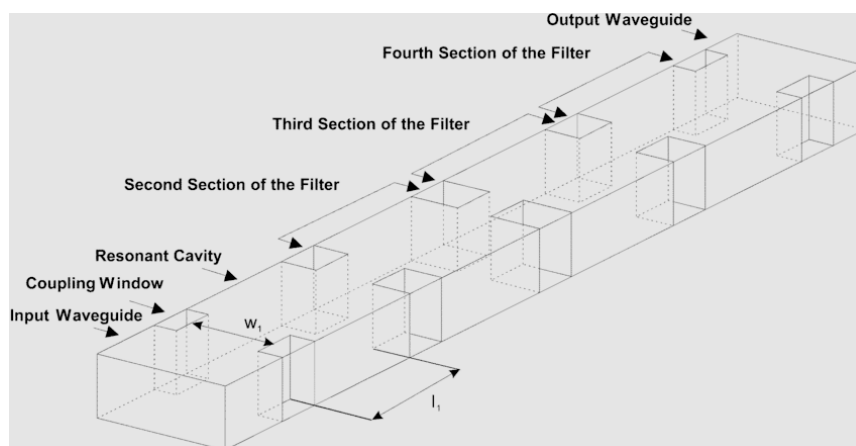
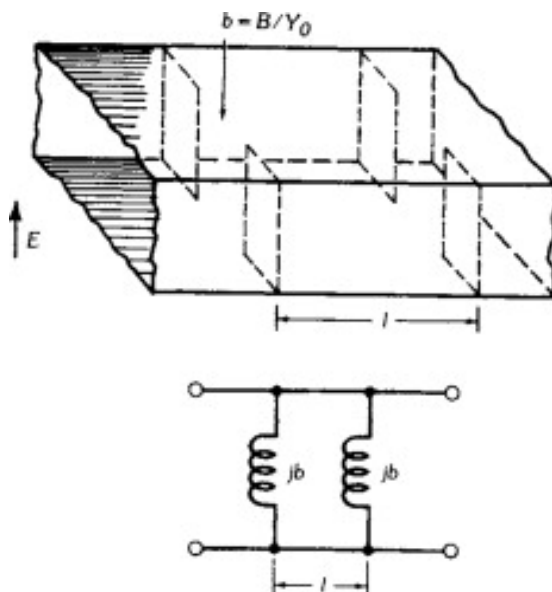
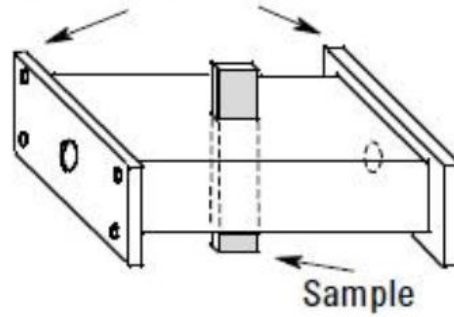
$$Q_c = \frac{2\omega W_e}{P_c}$$

$$P_d = \frac{1}{2} \int_V \sigma E^2 dV = \frac{\omega \epsilon''}{2} \int_V E^2 dV$$

$$Q_d = \frac{2\omega W_e}{P_d} = \frac{\epsilon'}{\epsilon''} = \frac{1}{\tan \delta}$$



Iris-coupled end plates



Lab

1. Design a rectangular cavity (TE₁₀₁ mode) resonator using WR-90 waveguide resonating at 10 GHz.
2. Design a circular coupling hole between a WR-90 rectangular waveguide and the above cavity resonator.
3. Simulate S₁₁. Plot S₁₁ (dB). Find the resonant frequency. Plot S₁₁ on the Smith chart.
4. Modify 2 into a through structure and simulate S₂₁. Plot S₂₁ (dB)